

Topic: Observability/Controllability

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1 Overview

- Controllability/Reachability is the property of a system concerning the ability to steer the state from arbitrary x_0 to arbitrary x_1 on a given time interval $[t_0, t_1]$.
- Observability is the property of a system concerning the ability to uniquely recover the initial state x_0 given the observation y .

1.1 Controllability/Reachability

For simplicity, let $D(\cdot) \equiv 0$ and consider our state space to be \mathbb{R}^n .

Definition 1 (Controllable). The system $\mathcal{D} = (A(\cdot), B(\cdot), C(\cdot))$ is controllable on $[t_0, t_1]$ if for all $(x_0, x_1) \in \mathbb{R}^n$, there exists $u_{[t_0, t_1]} \in \mathcal{U}$ which steers x_0 at t_0 to x_1 at t_1 .

We tend to break controllability into two different concepts:

- Controllability from the origin (reachability): the reachability map is given by

$$\mathcal{L}_r u(t) = \int_{t_0}^t \Phi(t, \tau) B(\tau) u(\tau) d\tau$$

If this map is surjective (i.e., $\text{Im}(\mathcal{L}_r) = \mathbb{R}^n$) then the system is reachable.

- Controllability to the origin (often simply referred to as controllability, hence, one should take care to identify the precise definition in the given reference they are looking at): the controllability map is given by

$$\mathcal{L}_c u(t) = \int_{t_0}^t \Phi(t_0, \tau) B(\tau) u(\tau) d\tau$$

If this map is surjective (i.e., $\text{Im}(\mathcal{L}_c) = \mathbb{R}^n$) then the system is controllable.

We can compute the adjoint of these two maps \mathcal{L}_r^* and \mathcal{L}_c^* respectively, and define the corresponding gram-mians

$$\mathcal{L}_r \mathcal{L}_r^* = \int_{t_0}^{t_1} \Phi(t_1, \tau) B(\tau) B^*(\tau) \Phi(t_1, \tau)^* d\tau$$

and

$$\mathcal{L}_c \mathcal{L}_c^* = \int_{t_0}^{t_1} \Phi(t_0, \tau) B(\tau) B^*(\tau) \Phi(t_0, \tau)^* d\tau$$

These are both linear operators and hence by the matrix representation theorem there is a finite dimension matrix representation $W_r \in \mathbb{R}^{n \times n}$ and $W_c \in \mathbb{R}^{n \times n}$, respectively. The finite rank operator lemma let's us show that

$$\text{Im}(\mathcal{L}_r) = \text{Im}(\mathcal{L}_r \mathcal{L}_r^*) = \text{Im}(W_r)$$

and

$$\text{Im}(\mathcal{L}_c) = \text{Im}(\mathcal{L}_c \mathcal{L}_c^*) = \text{Im}(W_c)$$

Moreover, we have the following equivalences:

$$\begin{aligned}
 (A(\cdot), B(\cdot)) \text{ controllable on } [t_0, t_1] &\iff \text{Im}(\mathcal{L}_r) = \mathbb{R}^n \\
 &\iff \text{Im}(\mathcal{L}_c) = \mathbb{R}^n \\
 &\iff \text{Im}(\mathcal{L}_r \mathcal{L}_r^*) = \mathbb{R}^n \\
 &\iff \text{Im}(\mathcal{L}_c \mathcal{L}_c^*) = \mathbb{R}^n \\
 &\iff \det(W_r) \neq 0 \\
 &\iff \det(W_c) \neq 0
 \end{aligned}$$

LTI Systems. Things get much easier in the LTI setting. Cayley Hamilton allows us to show that the condition on the controllability (reachability) grammian can be essentially reduced to checking a rank condition on the so called controllability matrix:

$$C = [B \quad AB \quad A^2B \quad \dots \quad A^{n-1}B] \in \mathbb{R}^{n \times nm}$$

Fact 2. The following equality holds:

$$\text{Im}(W_r) = \text{Im}(C)$$

Let $\Delta = t_1 - t_0$ for some $t_1 > t_0$.

Theorem 3. The following are equivalent:

$$\begin{aligned}
 &\text{The LTI system is completely controllable on some } [0, \Delta] \\
 \iff &\text{rank}([B \quad AB \quad \dots \quad A^{n-1}B]) = n && \text{(Rank Test)} \\
 \iff &\text{rank}([sI - A \quad B]) = n, \quad \forall s \in \mathbb{C} && \text{(PBH Test)}
 \end{aligned}$$

1.2 Observability

The pair $(A(\cdot), C(\cdot))$ is observable if given output $y(t)$, the initial state x_0 can be uniquely recovered.

Definition 4. The state x_0 is unobservable on $[t_0, t_1]$ if and only if its zero input response is zero on $[t_0, t_1]$.

Analogous to controllability, we can define the observability map $\mathcal{L}_o : \mathbb{R}^n \rightarrow \mathcal{Y}_{[t_0, t_1]}$ by

$$\mathcal{L}_o x_0(\cdot) = C(\cdot) \Phi(\cdot, t_0) x_0$$

That is, $\mathcal{L}_o x_0$ is an operator in $\text{PC}([t_0, t_1])$ such that

$$(\mathcal{L}_o x_0)(t) = y(t) = \int_{t_0}^t C(t) \Phi(t, \tau) B(\tau) u(\tau) d\tau$$

We have the following equivalences:

$$\begin{aligned}
 (A(\cdot), C(\cdot)) \text{ is completely observable (CO) on } [t_0, t_1] &\iff \text{Ker}(\mathcal{L}_o) = \{0\} \\
 &\iff \text{Ker}(\mathcal{L}_o^* \mathcal{L}_o) = \{0\} \\
 &\iff \det(W_o) \neq 0
 \end{aligned}$$

where

$$W_o = \int_{t_0}^{t_1} \Phi(\tau, t_0)^* C(\tau)^* C(\tau) \Phi(\tau, t_0) d\tau$$

LTI Systems. Just as with controllability, we can use Cayley Hamilton to construct an observability matrix:

$$\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{n-1} \end{bmatrix}$$

Theorem 5 (LTI Observability Tests). The following are equivalent:

The LTI system is completely observable on some $[0, \Delta]$

$$\Leftrightarrow \text{rank} \left(\begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} \right) = n \quad \text{(Rank Test)}$$

$$\Leftrightarrow \text{rank} \left(\begin{bmatrix} sI - A \\ C \end{bmatrix} \right) = n, \quad \forall s \in \mathbb{C} \quad \text{(PBH Test)}$$

2 Problems

We will start with some warm up problems.

Problem 1. (Controllability & Observability.) Consider the linear system given by

$$\begin{aligned} \dot{x} &= \begin{bmatrix} -3 & 3 \\ \gamma & -4 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \\ y &= [1 \quad 1] x \end{aligned}$$

for some parameter γ .

- How should we choose γ such that the system is controllable but not observable?
- How should we choose γ such that the system is observable but not controllable?

solution.

Problem 2. (Controllability & Observability.) Consider the linear system given by

$$\dot{x} = \begin{bmatrix} 1 & 0 \\ -1 & -2 \end{bmatrix} x$$

Suppose you have the ability to add one sensor and one actuator.

- a. Which state should we control with the actuator to make the system controllable?
- b. Which state should we measure with the sensor to make the system observable?

solution.

Problem 3. (Reachability.) Consider a unit point mass under control of force—i.e.,

$$\begin{bmatrix} \dot{x} \\ \dot{\dot{x}} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

This is equivalent to $\ddot{x} = u$. Suppose we want to reach $(x(T), \dot{x}(T)) = (1, 0)$ from $(x(0), \dot{x}(0)) = (0, 0)$ using a controller of the form

$$u(t) = \begin{cases} u_0, & 0 \leq t < T/10 \\ u_1, & T/10 \leq t < 2T/10 \\ \vdots & \vdots \\ u_{10}, & 9T/10 \leq t \leq T \end{cases}$$

Find $u = [u_1 \ u_2 \ \cdots \ u_{10}]^T$.

Problem 4. (Minimum Norm Control.) As we saw in the recorded lecture (and its worth re-emphasizing) that the controllability Grammian is related to the cost of control. Given $x(t_0) = x_0$, find a control function or sequence $u(\cdot)$ so that $x(t_1) = x_1$. Let $x_d = x_1 - \Phi(t_1, t_0)x_0$. Then we must have

$$x_d = \mathcal{L}_{r, [t_0, t_1]}(u)$$

where \mathcal{L}_r is the reachability map. Since we saw the continuous time version in the recorded lecture, for some variety let's focus on the discrete time case:

$$L_{r, [k_0, k_1]}(u) = \sum_{k=k_0}^{k_1-1} \Phi(k_1, k+1)B(k)u(k) = \mathcal{C}(k_0, k_1)U$$

where

$$\mathcal{C}(k_0, k_1) \in \mathbb{R}^{n \times (k_1 - k_0)m}, \quad \text{and } U = \begin{bmatrix} u(k_0) \\ \vdots \\ u(k_1 - 1) \end{bmatrix}$$

Since $x_d \in \text{Im}(L_{r, [k_0, k_1]})$ for solutions to exist, we assume $\text{Im}(L_{r, [k_0, k_1]}) = \mathbb{R}^n$. This implies that $W_{r, [k_0, k_1]}$ is invertible. Generally there are multiple solutions. To resolve the non-uniqueness, we find the solution so that

$$J(U) = \frac{1}{2} \sum_{k=k_0}^{k_1-1} u(k)^\top u(k) = U^\top U$$

is minimized.

Problem 5. (Pole Placement.) Consider

$$\dot{x} = \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

and the desired characteristic polynomial $p(s) = (s + 1)(s + 3)$. Design a feedback controller of the form $u = -kx$ to place the poles at those of $p(s)$.

solution.

Problem 6. (Connections between Observability & Lyapunov.) Let (A, C) be observable and suppose that P is any solution to $A^*P + PA = -C^*C$. Note that we are not assuming that $\mathcal{L}(\cdot)$ is an invertible operator here so there may be more than one solution. Show that $P \succeq 0$ if and only if A is stable.

solution.

Problem 7. (Controllable Canonical Form.) We saw in the recorded lecture and the lecture notes that if a system is completely controllable there is a transformation to the controllable canonical form. In many references this form is actually called the "controller canonical form" since its structure is invariant under the use of a feedback controller (for SISO systems this is easy to see). An alternative to this is what is called the "controllability canonical form" which is given by

$$\dot{x} = \underbrace{\begin{bmatrix} 0 & 0 & \cdots & 0 & -a_{n-1} \\ 1 & 0 & \cdots & 0 & -a_{n-2} \\ 0 & 1 & & 0 & -a_{n-3} \\ \vdots & & \ddots & & \vdots \\ 0 & 0 & & 1 & -a_0 \end{bmatrix}}_{A_c} x + \underbrace{\begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}}_{B_c} u$$

$$y = \underbrace{[c_1 \quad c_2 \quad c_3 \quad \cdots \quad c_n]}_{C_c} x$$

The controllability matrix for this system is the identity $C_c = I$ and (A_c, B_c) is of course controllable.

Consider another system (A, B) and suppose that it is controllable. Derive a transformation such that $x_c = T^{-1}x$ is in controllable form:

$$A_c = T^{-1}AT, \quad B_c = T^{-1}B$$

solution.